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Multiband MIMO Antenna by Mutually Coupling Non-radiating Edges for 4G

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Abstract

Wireless local area network (WLAN) technologies have been unified by means of multiband antenna. The challenge is to optimize the size of the multiband antenna for its inclusion in a small handheld terminal, while maintaining its performance characteristics. The challenge is further intensified as the multiple element antenna (MEA) are considered for compact diversity and multiple input multiple output (MIMO) terminal devices. MIMO is considered as a solution to overcome the problems of low data rates and maximize the channel bandwidth. Moreover, it also addresses the problems of multipath fading. Unfortunately, in real and robust environments there is a trade-off between number of MEA and MIMO performance. This is primarily due to the mutual coupling between the antennas. The proposed structure covers the 3G/4G range of 2.1 GHz and 2.5 GHz. The three configurations have multiband bands with $VSWR \leq 2$ are in the range of 950-2100 MHz and 2.4-3.7 GHz. The simulated and the measured results are in good agreement.

Keywords: Multiband; rectangular microstrip antenna; gap coupled rectangular microstrip antenna, multiple resonators; parasitic resonator; multiband MIMO antenna.

1. Introduction

MIMO (Multiple Input Multiple Output) systems are proved to achieve higher data rates by deploying multiple antennas at both the transmitter and receiver instead of a single antenna at the respective locations without using additional bandwidth or an increase in the respective power. MIMO systems are very much suitable for the present and emerging communications systems like Wi-Fi, 3G and 4G etc. However, when multiple antennas are involved at closer spacing the technical challenges are more pronounced compared to a SISO (Single Input Single Output) system. Hence, the basic aim of MIMO antenna design is to minimize the correlation between the multiple signals. The parameter that describes the correlation between the received signals in highly diversified environments is mutual coupling, which deteriorates the performance of the communications system. By calculating the mutual coupling, one can analyze the electromagnetic field interactions that exist between the elements of a MIMO system. Higher mutual coupling reduces the antenna efficiency and thus minimizes the system channel capacity. The impact of mutual coupling on the capacity of MIMO systems is studied. The mutual coupling mainly depends on the distance between the elements of an antenna array. By increasing the distance between the elements of the antennas, the mutual coupling can be reduced. However, the distance between the antennas cannot be maintained too large, since MIMO systems have their major applications in mobile terminals, laptops, and WLAN Access Points. Wireless communications. Patch antennas are very much compatible with MIMO systems because they are easier to fabricate and inexpensive, low in weight, planar or conformal layout. Patch antennas can be designed in any desired shape and this flexibility in patch antenna design makes it preferable for modern wireless communications. However, the patch antennas suffer from narrow bandwidth, which limits their application in modern communication systems like MIMO systems [1].

Multiple-input multiple-output (MIMO) technology, which is the key technology for the fourth generation mobile communication system (4G), has potentiality of increasing capacity without sacrificing additional spectrum.

Multiples antennas closely spaced will cause strong mutual coupling which deteriorates the performance of the MIMO system. Some various methods have been presented to increase the isolation of MIMO antenna, such as slit etched into the ground plane [2], an LC-based branch-line hybrid coupler [3] and neutralization line [4,5]. Unfortunately, all of those MIMO antennas could not cover the operation band of mobile phone for the third generation mobile communication system (3G). In order to get the operation band of mobile phone for 3G and take advantage of the MIMO technology, a multiband antenna, compact and a wideband MIMO antenna, is presented, and this antenna is suitable for mobile phone of 3G or fourth generation mobile phone (4G). A single RMSA is splitted into smaller elements along the width. One of the smaller elements is fed using co-axial probe while others are coupled to it's non-radiating edge [6,7]. The lengths and the gap between the elements is varied to increase the separation between different resonances to obtain dual and triple frequency operation. Configurations consisting a maximum till twelve strips have been investigated. Frequency ratio increases with increase in number of elements in the system. When the difference between the lengths of the different elements is larger then the separation between the resonances is more yielding larger frequency ratio, but the bandwidths at the individual frequency bands is less as matching is not optimum. As the number of elements increase, the flexibility for multi frequency operation also increases, as there are more number of elements that are varied to obtain the desired performance [8,9]. Configuration from ten elements to twelve elements have been obtained which yields four and five individual frequency bands with sufficient separation between the individual bands has been presented in this paper. Also, as bandwidth increases with increase in number of elements, considerably wide bandwidth can be obtained at individual resonance frequencies. However, as the width of individual element decreases with the increase in number of elements, gain and efficiency also decrease [10]. So these gap coupled configurations suffer from poor gain and may not be suitable for applications with high gain requirements. So, techniques can be devised to improve the gain of these configurations [11,12].

2. Multiband Configuration

With increase in number of elements, number of gaps increase, and so their optimization becomes quite tedious. So detailed observations for the effect of gap size on frequency ratio are made, which may serve as an aid for effective design of these gap coupled configurations. As regards the effect of feed probe diameter, it has been observed that with increase in feed probe diameter, the bandwidth increases but does not have significant effect on the frequency ratio. This is because the inductance of probe decreases with increase in diameter, thereby yielding higher bandwidth due to improved matching. Also, it has been found that the bandwidth increases with increase in loss tangent, at the cost of efficiency. The radiation pattern of these configurations has also been studied in detail. It has been found that radiation pattern of configurations with odd number of elements is more regular, as compared to the configurations with even number of elements and remains more or less in broad side direction at all the frequencies.

The Four Band configuration is obtained with ten elements with frequencies as listed in Table I.

TABLE I FOUR FREQUENCY RESPONSE OF CONFIGURATION WITH TEN STRIPS (Cr = 4.3, h = 1.59 mm, tan δ = 0.02, W = 4.18 mm)

Lengths L (mm)	Gaps, S (mm)	x (mm)	fr1 (GHz)	2.072
			RL1 (dB)	-17.8
			BW1 (MHz)	61
L1=31	S1=0.575	8	fr2 (GHz)	2.304
L2=33	S2=0.425		RL2 (dB)	-26.5
L3=34	S3=0.575		BW2 (MHz)	60
L4= 32	S4=0.2		fr3 (GHz)	2.456
L5 (fed)=33	S5=0.2		RL3 (dB)	-30.59
L6= 36	S6=0.45		BW3 (MHz)	81
L7=34	S7=0.45		fr4 (GHz)	2.637
L8= 33	S8=0.475		RL4 (dB)	-16.08
L9= 31	S9=0.475			
L10=32			BW4 (MHz)	52

the frequencies are able to be in the range for 3G and 4G mobile applications. Also the mutual coupling between the frequencies is reduced as shown in Fig. 1. The bandwidths for the individual frequency bands is almost same with considerable isolation between the individual frequency bands is achieved with this configuration. The First

frequency is at 2.072GHz and the fourth at 2.637GHz.

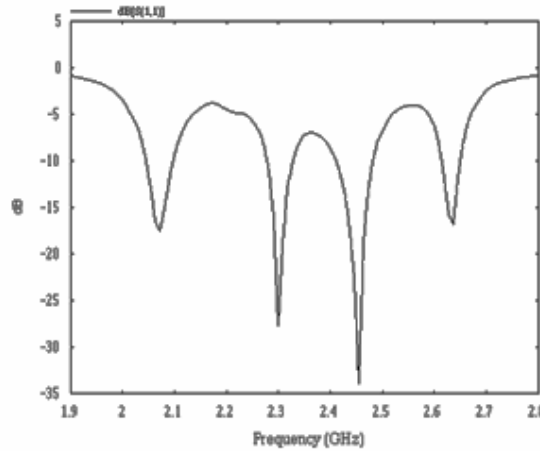


Figure 1. Four frequency response of ten patch configuration S11 plot

Now the single element is splitted into eleven elements the width of the individual element is again kept constant and the length is varied along with the gap coupling. The results are illustrated in Table II. The first resonant frequency is same as that of the ten band configuration but the fourth individual frequency has changed to 2.645GHz. The bandwidths are also improved for the third frequency band whereas it is same for the fourth frequency band. This configuration has been fabricated using FR4 substrate with parameters, $h = 1.59\text{mm}$, $C_r = 4.3$, and $\tan\delta = 0.02$.

TABLE II- FOUR FREQUENCY RESPONSE OF CONFIGURATION WITH ELEVEN STRIPS ($C_r = 4.3$, $h = 1.59\text{mm}$, $\tan\delta = 0.02$, $W = 4.18\text{mm}$)

Lengths L (mm)	Gaps, S (mm)	X (mm)	fr1 (GHz)	2.027
			RL1 (dB)	-18.62
$L_1=30$ $L_2=34$ $L_3=36$ $L_4=36$ $L_5=36$ $L_{6(\text{fed})}=33$ $L_7=31$ $L_8=30$ $L_9=32$ $L_{10}=34$ $L_{11}=31$	$S_1=1.475$ $S_2=0.35$ $S_3=0.7$ $S_4=1.05$ $S_5=0.2$ $S_6=0.2$ $S_7=1.05$ $S_8=0.7$ $S_9=1.225$ $S_{10}=1$	9	BW1 (MHz)	49
			fr2 (GHz)	2.227
			RL2 (dB)	-12.87
			BW2 (MHz)	64
			fr3 (GHz)	2.364
			RL3 (dB)	-16.47
			BW3 (MHz)	112
			fr4 (GHz)	2.473
			RL4 (dB)	-24.74
			BW4 (MHz)	73
			fr5 (GHz)	2.645
RL5 (dB)	-17.21			
BW5 (MHz)	52			

Fig. 2 shows the experimental and simulated results of return loss in dB versus frequency in GHz for eleven elements that yield four individual frequency bands with increased isolation between the individual frequency bands.

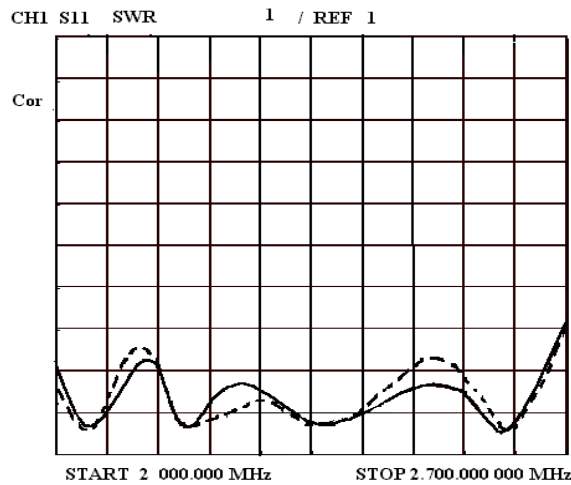


Figure 2. Comparative plot for measured (—) and simulated (-----) VSWR for eleven patch configuration.

The further addition for twelve element considerably increases the separation between individual frequency bands, Fig. 3 shows the comparative plot of return loss for eleven and twelve strip configuration for four frequency bands operation. The fourth frequency band obtained with twelve elements is 2.88 GHz. The eleven and twelve patch configuration show the four frequency band operation but due to increase in elements show the frequency separation ratio for first two frequencies with eleven patches is 1.07 whereas it is 1.09 for twelve patches.

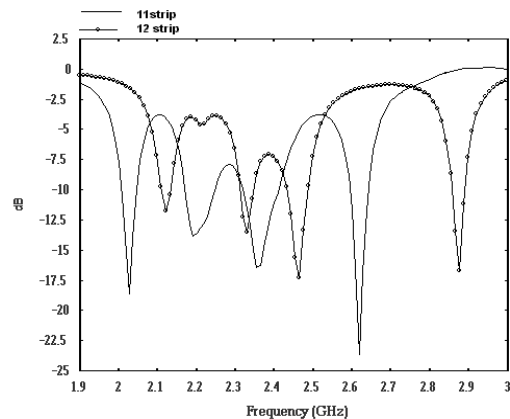


Figure 3. Comparative plot of return loss for eleven and twelve strip configuration for four frequency bands operation.

For separation between second and third frequency bands is same for both the configurations. Now the separation between the third and fourth frequency band has drastically increased with twelve patches as compared to eleven patches from 1.10 to 1.16.

The optimization for five frequency bands with ten and twelve elements was unsuccessful. The five individual frequency bands were achieved only with the eleven elements configuration. The optimization for five frequency bands did not increase the frequency ratio. But this configuration has five distinctive individual frequencies with return loss that is acceptable. With many even number of elements the optimization is not possible as the distribution of elements with respect to the centre feed element is unequal so the resonant frequencies are not optimized with considerable separation between the individual frequencies. The results for five frequencies with the eleven elements is specified in Table III. The optimization with further addition of elements does not show improvement in the increase of number of frequency bands due to increase of mutual coupling between the individual elements. Mutual coupling has been reduced in the above configuration and thus the multibands were successfully simulated and measured.

TABLE III-FIVE FREQUENCY RESPONSE OF CONFIGURATION WITH ELEVEN STRIPS ($\epsilon_r = 4.3$, $h = 1.59$ mm, $\tan\delta = 0.02$, $W = 4.18$ mm)

<i>Lengths L (mm)</i>	<i>Gaps, S (mm)</i>	<i>x (mm)</i>	fr1 (GHz)	2.027
			RL1 (dB)	-18.62
$L_1=30$ $L_2=34$ $L_3=36$ $L_4=36$ $L_5=36$ $L_{6(\text{fed})}=33$ $L_7=31$ $L_8=30$ $L_9=32$ $L_{10}=34$ $L_{11}=31$	$S_1=1.475$ $S_2=0.35$ $S_3=0.7$ $S_4=1.05$ $S_5=0.2$ $S_6=0.2$ $S_7=1.05$ $S_8=0.7$ $S_9=1.225$ $S_{10}=1$	9	BW1 (MHz)	49
			fr2 (GHz)	2.227
			RL2 (dB)	-12.87
			BW2 (MHz)	64
			fr3 (GHz)	2.364
			RL3 (dB)	-16.47
			BW3 (MHz)	112
			fr4 (GHz)	2.473
			RL4 (dB)	-24.74
			BW4 (MHz)	73
			fr5 (GHz)	2.645
			RL5 (dB)	-17.21
			BW5 (MHz)	52

3. Conclusion

In this paper multiple frequencies up to five bands are obtained by dividing a single resonator along the width so that the width of each element is equal, then the lengths and gap between the individual elements is varied. The mutual coupling between the elements is reduced that yield multiple frequencies. But there is limitation on the number of elements as the mutual coupling increases that combines the individual frequency bands. The configuration with ten, eleven and twelve structure covers the 3G/4G range of 2.1 GHz and 2.5 GHz. These separation between the individual frequency. The three configurations have multiband bands with $VSWR \leq 2$ are in the range of 950–2100 MHz and 2.4–3.7 GHz. These characteristics are well suited for all 4G MIMO applications.

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